

Low Noise Bias Generator in 2mm × 2mm DFN

FEATURES

- Generates Low Noise Bias Voltage from Single Cell Li-Ion Battery
- Output Current: Up to 40mA
- Low Output Ripple and Noise: <500µV_{P-P}
- Boost Regulator
 - 1MHz Operation (LT3048)
 - 2.2MHz Operation (LT3048-XX)
 - Integrated Schottky Diode
- Low Dropout Linear Regulator
 - 1.235V Feedback Voltage (LT3048)
 - Low Noise: <120µV_{RMS} (10Hz to 100kHz)
 - 0.1% Load Regulation
- Fixed 3.3V, 5V, 12V, 15V and Adjustable Output Options
- Short-Circuit and Thermal Protection
- Load Disconnect in Shutdown
- Available in 2mm × 2mm DFN Package

APPLICATIONS

- Sensor Bias
- Op Amp Supply

DESCRIPTION

The LT®3048 series generate low noise, low ripple bias supplies from input voltages of 2.7V to 4.8V.

The LT3048 includes a boost regulator and a LD0 linear regulator. The boost regulator provides power to the linear regulator.

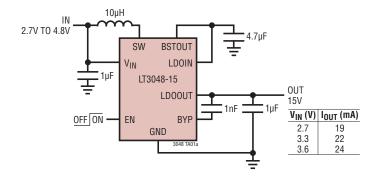
The boost regulator output voltage is regulated to 1.1V above the LDO output, optimizing LDO ripple rejection and transient response. Fixed frequency operation and current mode control allow the use of very small inductors and results in low, predictable output ripple.

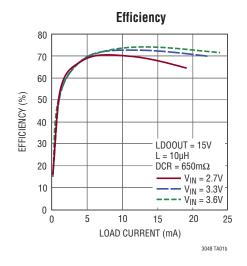
The linear regulator in the LT3048 generates a programmable output and the LT3048-XX generate fixed output voltages. High power supply ripple rejection combined with a low noise internal reference results in less than $500\mu V_{P-P}$ output ripple and noise.

The LT3048 is available in a 8-lead 2mm \times 2mm DFN package.

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TYPICAL APPLICATION





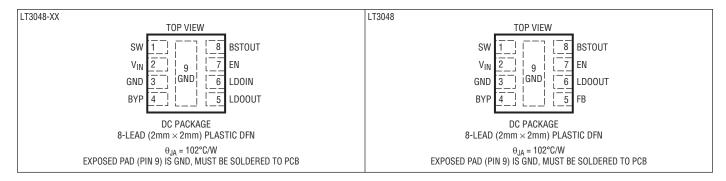
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ABSOLUTE MAXIMUM RATINGS

(וזטנט ו)	
V _{IN}	6V
SW	25V
BSTOUT	23V
LDOIN	22V
LD00UT	22V
FB	6V
BYP	

Operating Junction Temperature Range	e (Note 2)
E-Grade	–40°C to 125°C
I-Grade	–40°C to 125°C
Maximum Junction Temperature	125°C
Storage Temperature Range	–65°C to 150°C

PIN CONFIGURATION



ORDER INFORMATION

LEAD FREE FINISH	TAPE AND REEL	PART MARKING	PACKAGE DESCRIPTION	TEMPERATURE RANGE
LT3048EDC#PBF	LT3048EDC#TRPBF	LGRR	8-Lead (2mm × 2mm) Plastic DFN	-40°C to 125°C
LT3048IDC#PBF	LT3048IDC#TRPBF	LGRR	8-Lead (2mm × 2mm) Plastic DFN	-40°C to 125°C
LT3048EDC-15#PBF	LT3048EDC-15#TRPBF	LGKK	8-Lead (2mm × 2mm) Plastic DFN	-40°C to 125°C
LT3048IDC-15#PBF	LT3048IDC-15#TRPBF	LGKK	8-Lead (2mm × 2mm) Plastic DFN	-40°C to 125°C
LT3048EDC-12#PBF	LT3048EDC-12#TRPBF	LGQQ	8-Lead (2mm × 2mm) Plastic DFN	-40°C to 125°C
LT3048IDC-12#PBF	LT3048IDC-12#TRPBF	LGQQ	8-Lead (2mm × 2mm) Plastic DFN	-40°C to 125°C
LT3048EDC-5#PBF	LT3048EDC-5#TRPBF	LGQR	8-Lead (2mm × 2mm) Plastic DFN	-40°C to 125°C
LT3048IDC-5#PBF	LT3048IDC-5#TRPBF	LGQR	8-Lead (2mm × 2mm) Plastic DFN	-40°C to 125°C
LT3048EDC-3.3#PBF	LT3048EDC-3.3#TRPBF	LGQS	8-Lead (2mm × 2mm) Plastic DFN	-40°C to 125°C
LT3048IDC-3.3t#PBF	LT3048IDC-3.3#TRPBF	LGQS	8-Lead (2mm × 2mm) Plastic DFN	-40°C to 125°C

Consult LTC Marketing for parts specified with wider operating temperature ranges.

Consult LTC Marketing for information on nonstandard lead based finish parts.

For more information on lead free part marking, go to: http://www.linear.com/leadfree/For more information on tape and reel specifications, go to: http://www.linear.com/tapeandreel/

ELECTRICAL CHARACTERISTICS The \bullet denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^{\circ}C$. $V_{IN} = 3.6V$, EN/UVLO = 3V, unless otherwise noted. Parameter applies to all options unless specified.

PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
V _{IN} Operating Range			2.7		4.8	V
Boost Regulator						
BSTOUT-LDOOUT Regulation Voltage		•	1	1.1	1.2	V
BSTOUT-V _{IN} Regulation Voltage	LD00UT < V _{IN}			1.1		V
V _{IN} Quiescent Current	Not Switching EN ≤ 0.4V			475 0.1	600 1	μΑ μΑ
Switching Frequency	LT3048	•	0.75 0.7	1.00 1.00	1.25 1.30	MHz MHz
	LT3048-XX	•	1.7 1.6	2.2 2.2	2.7 2.8	MHz MHz
Maximum Duty Cycle	LT3048 LT3048-XX		93 86	95 90		% %
Switch Current Limit	(Note 3)		300	400	500	mA
Switch V _{CESAT}	I _{SW} = 200mA			220		mV
Diode Forward Drop	I _{DIODE} = 100mA			0.84		V
EN Voltage High			1.25			V
EN Voltage Low					0.4	V
LDO Linear Regulator						
LDO Quiescent Current	LT3048: Supplied by LDOIN			14	20	μA
	LT3048-XX: Supplied by BSTOUT			14	20	μA
FB Pin Current				60		nA
FB Regulation Voltage	LT3048: I_{LOAD} = 100 μ A, FB = LD00UT LT3048: 100 μ A \leq I_{LOAD} \leq 40mA	•	1.216 1.210	1.235 1.235	1.254 1.260	V
FB Load Regulation	LT3048: ΔI_{LOAD} = 100 μ A to 40mA			2	6	mV
LDOOUT Regulation Voltage	LT3048-15: $I_{LOAD} = 100\mu A$ LT3048-15: $100\mu A \le I_{LOAD} \le 40mA$	•	14.775 14.625	15.00 15.00	15.225 15.375	V
	LT3048-12: $I_{LOAD} = 100\mu A$ LT3048-12: $100\mu A \le I_{LOAD} \le 40mA$	•	11.82 11.70	12.00 12.00	12.18 12.30	V
	LT3048-5: $I_{LOAD} = 100\mu A$ LT3048-5: $100\mu A \le I_{LOAD} \le 40mA$	•	4.925 4.875	5.00 5.00	5.075 5.125	V
	LT3048-3.3: $I_{LOAD} = 100\mu A$ LT3048-3.3: $100\mu A \le I_{LOAD} \le 40mA$	•	3.250 3.217	3.300 3.300	3.350 3.383	V
LD00UT Load Regulation	LT3048-15: ΔI _{LOAD} = 100μA to 40mA			20	45	mV
	LT3048-12: ΔI _{LOAD} = 100μA to 40mA			20	45	mV
	LT3048-5: ΔI _{LOAD} = 100μA to 40mA			12	25	mV
	LT3048-3.3: ΔI _{LOAD} = 100μA to 40mA			9	18	mV



ELECTRICAL CHARACTERISTICS The \bullet denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^{\circ}C$. $V_{IN} = 3.6V$, EN/UVLO = 3V, unless otherwise noted. Parameter applies to all options unless specified.

PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
LDOOUT Voltage Noise BW = 10Hz to 100kHz	LT3048: C _{BYP} = 0nF, I _{LOAD} = 40mA, LD00UT = 1.235V LT3048: C _{BYP} = 1nF, I _{LOAD} = 40mA, LD00UT = 1.235V			38 25		μV _{RMS} μV _{RMS}
$C_{OUT} = 1\mu F$	LT3048-15: $C_{BYP} = 0nF$, $I_{LOAD} = 24mA$ LT3048-15: $C_{BYP} = 1nF$, $I_{LOAD} = 24mA$			250 60		μV _{RMS} μV _{RMS}
	LT3048-12: $C_{BYP} = 0nF$, $I_{LOAD} = 31mA$ LT3048-12: $C_{BYP} = 1nF$, $I_{LOAD} = 31mA$			225 60		μV _{RMS} μV _{RMS}
	LT3048-5: C _{BYP} = 0nF, I _{LOAD} = 40mA LT3048:-5: C _{BYP} = 1nF, I _{LOAD} = 40mA			145 50		μV _{RMS} μV _{RMS}
	LT3048-3.3: $C_{BYP} = 0nF$, $I_{LOAD} = 40mA$ LT3048-3.3: $C_{BYP} = 1nF$, $I_{LOAD} = 40mA$			120 42		μV _{RMS} μV _{RMS}
Ripple Rejection	LT3048: f _{RIPPLE} = 1.0MHz, I _{LOAD} = 40mA LT3048: f _{RIPPLE} = 2.2MHz, I _{LOAD} = 40mA			60 53		dB dB
Current Limit		•	45	80		mA

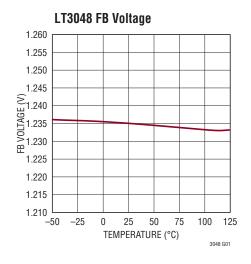
Note 1: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

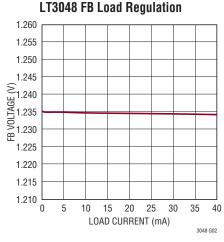
Note 2: The LT3048E is guaranteed to meet performance specifications from 0°C to 125°C junction temperature. Specifications over the -40°C to 125°C operating junction temperature range are assured by design, characterization and correlation with statistical process controls. The

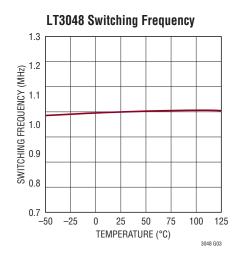
LT3048I is guaranteed over the full –40°C to 125°C operating junction temperature range. High junction temperatures degrade operating lifetimes. Operating lifetime is derated at junction temperatures greater than 125°C.

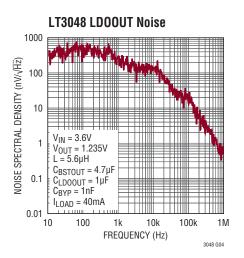
Note 3: Current limit guaranteed by design and/or correlation to static test. Slope compensation reduces current limit at higher duty cycle.

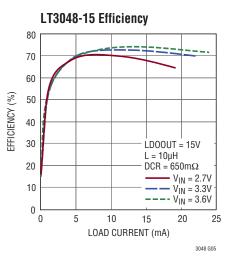
TYPICAL PERFORMANCE CHARACTERISTICS

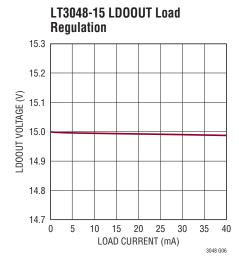


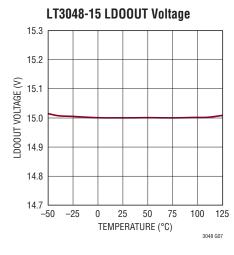


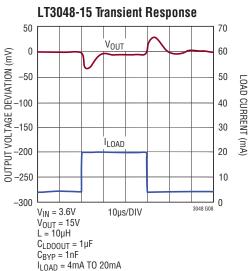


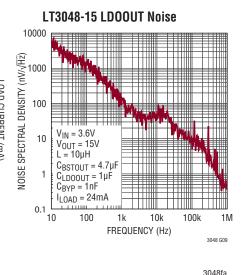




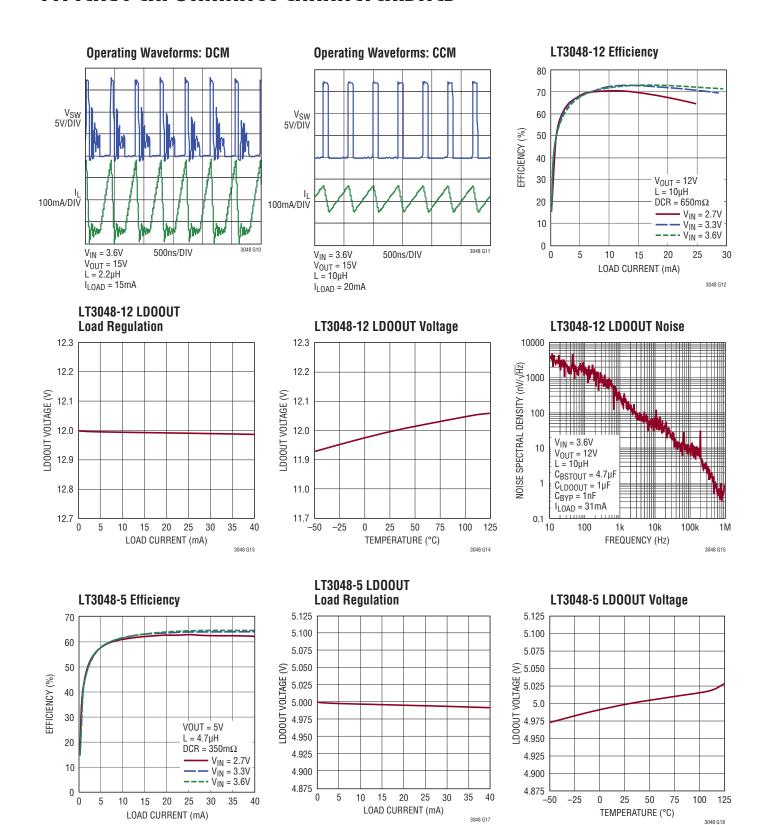








TYPICAL PERFORMANCE CHARACTERISTICS

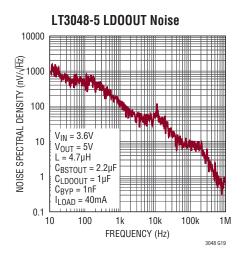


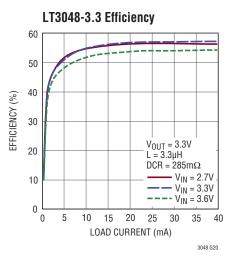
LINEAR

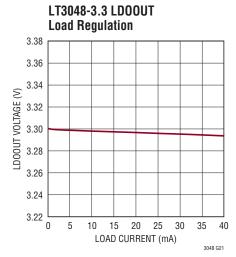
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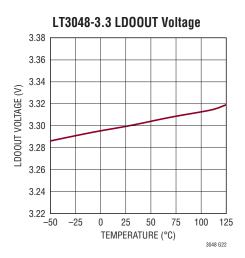
3048 G16

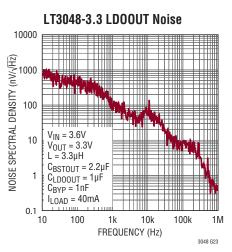
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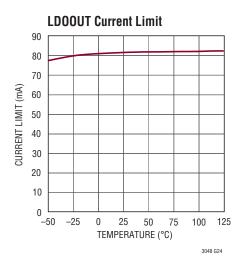


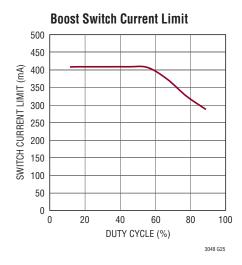


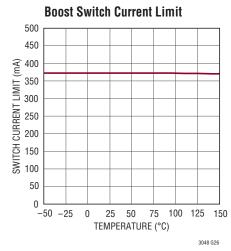


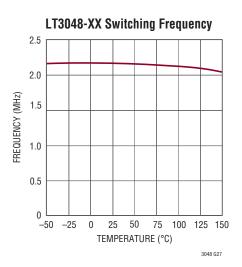












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PIN FUNCTIONS

LT3048

SW (**Pin 1**): Boost Regulator Switch Node. This is the collector of the internal power transistor and the anode of the internal Schottky diode.

V_{IN} (**Pin 2**): Input Supply. Provides operating current to the boost regulator. Bypass to ground.

GND (Pin 3): Ground. This pin must be soldered to PCB ground. The exposed pad must also be soldered to PCB ground.

BYP (Pin 4): Bypass. Add an optional capacitor from LDOOUT to BYP to reduce noise at LDOOUT; otherwise leave BYP unconnected. A capacitor also soft-starts the linear regulator.

FB (Pin 5): The LT3048 regulates the FB pin to 1.25V. Connect the feedback resistor divider tap to this pin.

LDOOUT (Pin 6): LDO Linear Regulator Output. This pin is programmed by the feedback resistor divider. Bypass to ground with a minimum of 1µF.

EN (Pin 7): Enable. Logic input to enable the boost and linear regulator.

BSTOUT (Pin 8): Boost Regulator Output. This pin is connected to the internal Schottky diode and is regulated to 1.1V above LDOOUT. Bypass to ground.

GND (Exposed Pad Pin 9): The internal boost power switch is connected to the exposed pad. Tie to ground.

LT3048-XX

SW (Pin 1): Boost Regulator Switch Node. This is the collector of the internal power transistor and the anode of the internal Schottky diode.

V_{IN} (**Pin 2**): Input Supply. Provides operating current to the boost regulator. Bypass to ground.

GND (Pin 3): Ground. This pin must be soldered to PCB ground. The exposed pad must also be soldered to PCB ground.

BYP (Pin 4): Bypass. Add an optional capacitor from LDOOUT to BYP to reduce noise at LDOOUT; otherwise leave BYP unconnected. A capacitor also soft-starts the linear regulator.

LDOOUT (Pin 5): LDO Linear Regulator Output. This pin is regulated to a fixed voltage. Bypass to ground with a minimum of $1\mu F$.

LDOIN (Pin 6): LDO Linear Regulator Input. Normally connected to BSTOUT. Bypass to ground.

EN (Pin 7): Enable. Logic input to enable the boost and linear regulator.

BSTOUT (Pin 8): Boost Regulator Output. This pin is connected to the internal Schottky diode and is regulated to 1.1V above LDOOUT. Bypass to ground.

GND (Exposed Pad Pin 9): The internal boost power switch is connected to the exposed pad. Tie to ground.



LT3048 BLOCK DIAGRAMS

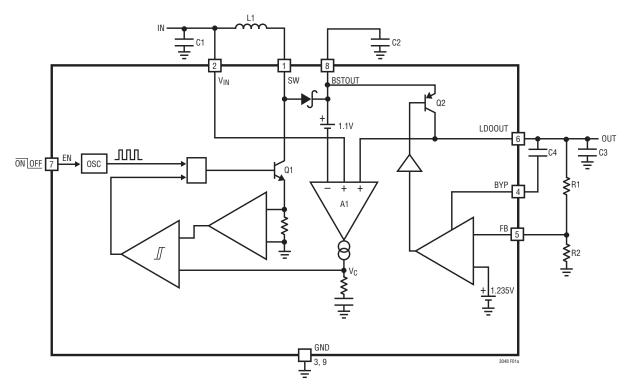


Figure 1a. LT3048 Block Diagram

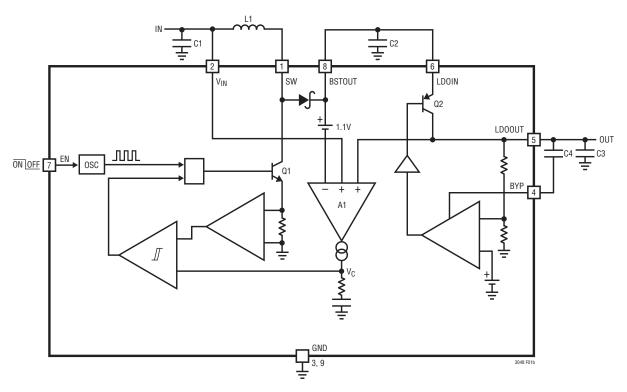


Figure 1b. LT3048-XX Block Diagram



OPERATION

The LT3048 combines a boost regulator with a low dropout linear regulator to produce a programmable output from a lower input voltage. The LT3048-15, LT3048-12, LT3048-5, LT3048-3.3 regulate to 15V, 12V, 5V and 3.3V, respectively. With EN below 0.4V, all circuits are turned off, Q2 disconnects the load from the input, and current consumption is less than $1\mu A$. Driving EN high enables the oscillator and all bias circuits including voltage reference and regulation amplifiers.

The boost regulator uses peak current mode operation, providing cycle-by-cycle current regulation and limiting,

fast transient response, and good stability. The error amplifier, A1, regulates BSTOUT (and LDOIN) to 1.1V above LDOOUT or $V_{\rm IN}$, whichever is higher.

Regulating 1.1V between LDOIN and LDOOUT allows the linear regulator to provide good transient response and ripple rejection, while maintaining good efficiency. This regulation loop also minimizes total power dissipation in fault conditions. If the output is overloaded, the linear regulator will enter current limit, LDOOUT voltage will decrease. By reducing the boost regulator output voltage, the voltage across Q2 will be reduced, limiting dissipation in Q2. Thermal shutdown provides additional protection.

APPLICATIONS INFORMATION

Capacitor Selection

A $4.7\mu F$ ceramic capacitor should be sufficient for the boost regulator output bypass. Low ESR (equivalent series resistance) ceramic capacitors should be used at the outputs of the regulator to minimize the output ripple voltage. Use only X5R or X7R dielectrics, as these materials retain their capacitance over wider voltage and temperature ranges better than other dielectrics.

Table 1. Ceramic Capacitor Manufacturers

MANUFACTURER	PHONE	WEB
Taiyo Yuden	(408) 573-4150	www.t-yuden.com
AVX	(803) 448-9411	www.avxcorp.com
Murata	(714) 852-2001	www.murata.com

FB Resistor Network

The output voltage is programmed with a resistor divider between the output and FB pin. Choose the 1% resistors according to:

$$R1 = R2 \left(\frac{V_{0UT}}{1.235V} - 1 \right)$$

R2 should be 200k or less to avoid bias current errors.

Inductor Selection

A 5.6µH inductor will suffice for most LT3048 applications. See Table 2 for recommended inductor values. Smaller value inductors can be used at the cost of reduced maximum load current. Figure 2 shows the maximum load current as a function of output voltage. The inductor's RMS current rating must be greater than the maximum input current. To keep efficiency high, the DCR (series resistance) should be minimized. Table 3 lists several vendors.

Table 2. Recommended Inductor Values

PART #	INDUCTANCE
LT3048	5.6µH
LT3048-15	10μH
LT3048-12	10μH
LT3048-5	4.7μΗ
LT3048-3.3	3.3µH

Table 3. Inductor Manufacturers

VENDOR	URL
Coilcraft	www.coilcraft.com
Sumida	www.sumida.com
Toko	www.toko.com
Würth Elektronik	www.we-online.com

LINEAR TECHNOLOGY

APPLICATIONS INFORMATION

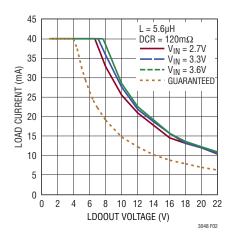


Figure 2. LT3048 Typical and Guaranteed Load Current vs. Output Voltage

Reducing Output Noise With A Bypass Capacitor

The LT3048 relies on the power supply rejection of the linear regulator to reduce switching regulator noise at LD00UT. The linear regulator also contributes thermal noise to the output. The thermal noise can be reduced, and transient response improved, by adding a capacitor between LD00UT and BYP. A typical value is 1nF. This capacitor increases start-up time of the regulator.

Recommended PCB Layout

Figure 3 shows the recommended layout for LT3048 circuits. Most important is careful placement of the BSTOUT bypass capacitor C2. High frequency AC current flows in a loop formed by C2, internal power transistor Q1 and boost diode D2. Keep this loop small. Also be sure to place an unbroken ground plane below this loop, on the highest copper layer below the surface. This prevents the AC loop from coupling to LDOOUT and other nearby circuitry. Keep the SW node as small as possible.

LDO Stability and Output Capacitance

The LT3048 linear regulator requires an output capacitor for stability. It is designed to be stable with most low ESR capacitors (typically ceramic, tantalum or low ESR electrolytic). A minimum output capacitor of $1\mu F$ with an ESR of 1Ω or less is recommended to prevent oscillations. Larger values of output capacitance decrease peak

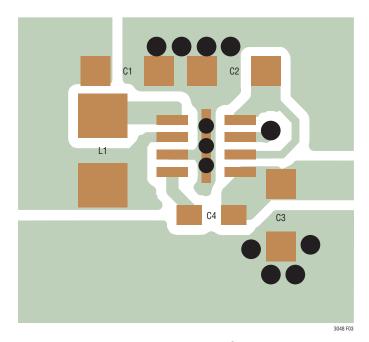


Figure 3. Recommended PCB Layout

deviations and provide improved transient response for larger load current changes. Bypass capacitors, used to decouple individual components powered by the LT3048, increase the effective output capacitor value.

Give consideration to the use of ceramic capacitors as they are manufactured with a variety of dielectrics, each with different behavior across temperature and applied voltage. The most common dielectrics used are specified with EIA codes of Z5U, Y5V, X5R and X7R. Typical voltage and temperature coefficients are shown in Figures 4 and 5. The X5R and X7R dielectrics have more stable characteristics and are most suitable for use as the output capacitor. The X7R type has better stability across temperature, while the X5R is less expensive and is available in higher values. Care still must be exercised when using X5R and X7R capacitors; the codes only specify operating temperature range and maximum capacitance change over temperature. Capacitance change due to DC bias with X5R and X7R capacitors can be significant enough to drop capacitor values below appropriate levels. Capacitor DC bias characteristics tend to improve as case size increases, but expected capacitance at operating voltage should be verified.



APPLICATIONS INFORMATION

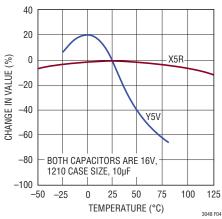


Figure 4. Ceramic Capacitor Temperature Characteristics

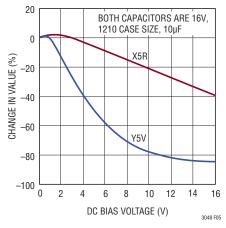


Figure 5. Ceramic Capacitor DC Bias Characteristics

Voltage and temperature coefficients are not the only sources of problems. Some ceramic capacitors have a piezoelectric response. A piezoelectric device generates voltage across its terminals due to mechanical stress. In a ceramic capacitor, the stress can be induced by vibrations in the system or thermal transients.

Wide Input Voltage Applications

The LT3048 can be used in wide input voltage applications. The input supply voltage to the LT3048 must be between 2.7V and 4.8V and capable of at least 10mA. The inductor can run off a separate voltage supply (PV_{IN}). This technique allows the output to be powered from 1V to 23V (see Figure 6).

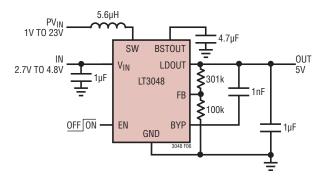
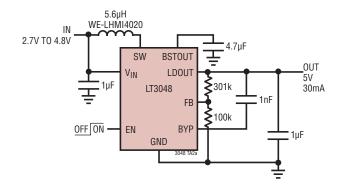


Figure 6. Connecting the Inductor to a Separate Voltage Supply

APPLICATIONS INFORMATION

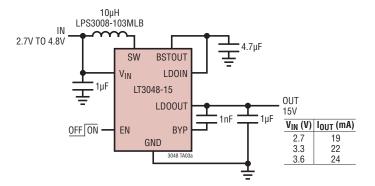
LT3048



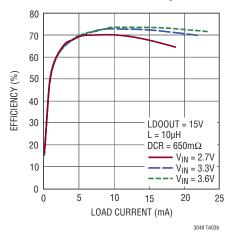
70 60 50 EFFICIENCY (%) 40 30 VOUT = 5V $L = 5.6 \mu H$ 20 $DCR = 120m\Omega$ $V_{IN} = 2.7V$ 10 $-V_{IN} = 3.3V$ --- V_{IN} = 3.6V 0 5 20 25 0 10 15 30 LOAD CURRENT (mA) 3048 TA02b

LT3048 Efficiency

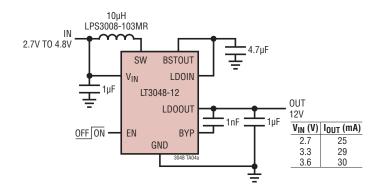
LT3048-15



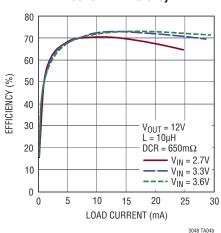




LT3048-12



LT3048-12: Efficiency

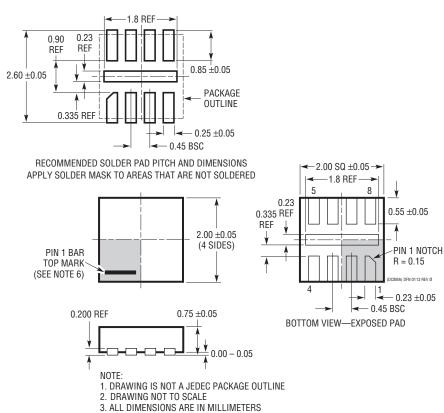


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PACKAGE DESCRIPTION

Please refer to http://www.linear.com/designtools/packaging/ for the most recent package drawings.

DC8 Package 8-Lead Plastic DFN (2mm × 2mm) (Reference LTC DWG # 05-08-1939 Rev Ø) **Exposed Pad Variation AA**



- 4. DIMENSIONS OF EXPOSED PAD ON BOTTOM OF PACKAGE DO NOT INCLUDE MOLD FLASH. MOLD FLASH, IF PRESENT, SHALL NOT EXCEED 0.15mm ON ANY SIDE
- 5. EXPOSED PAD SHALL BE SOLDER PLATED
- 6. SHADED AREA IS ONLY A REFERENCE FOR PIN 1 LOCATION ON THE TOP AND BOTTOM OF PACKAGE

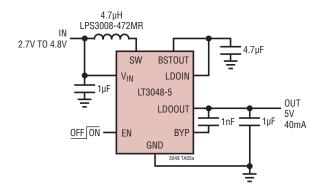
REVISION HISTORY

REV	DATE	DESCRIPTION	PAGE NUMBER
Α	2/15	Clarified to Include All Fixed and Adjustable Output Options	1 – 16

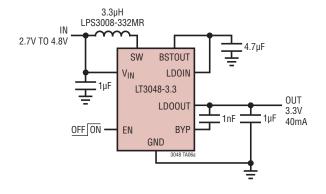


TYPICAL APPLICATIONS

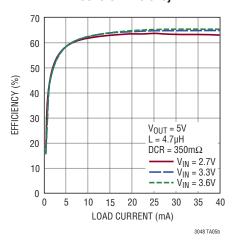
LT3048-5



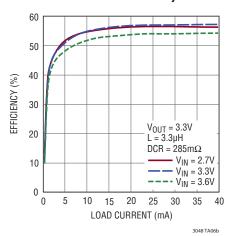
LT3048-3.3



LT3048-5: Efficiency



LT3048-3.3: Efficiency



RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
LT3460	300mA, 38V I _{SW} , 1.3MHz, High Efficiency Step-Up DC/DC Converter	$V_{IN}\!\!: 2.5V$ to 16V, $V_{OUT(MAX)}$ = 36V, I_Q = 2mA, I_{SD} < 1 μ A, SC70, ThinSOTTM Package
LT3461/LT3461A	300mA, 38V I _{SW} , 1.3MHz, High Efficiency Step-Up DC/DC Converters	$V_{IN}\!\!: 2.5V$ to 16V, $V_{OUT(MAX)}$ = 38V, I_Q = 2.8mA, I_{SD} < 1 μ A, SC70, ThinSOT Packages
LT3464	120mA, 38V I _{SW} , High Efficiency Step-Up DC/DC Converter with Integrated Schottky, Output Disconnect	V_{IN} : 2.3V to 10V, $V_{\text{OUT}(\text{MAX})}$ = 34V, I_{Q} = 25µA, I_{SD} < 1µA, ThinSOT Package
LT1613	550mA I _{SW} , 1.4MHz, High Efficiency Step-Up DC/DC Converter	$V_{\text{IN}} \!\!: 0.9 V$ to 10V, $V_{\text{OUT}(\text{MAX})} = 34 V$, $I_{\text{Q}} = 3 \text{mA}$, $I_{\text{SD}} < 1 \mu \text{A}$, ThinSOT Package
LT1761	20V, 100mA (I _{OUT}) Low Noise LDO	V_{IN} : 1.8V to 20V, $V_{\text{OUT}(\text{MIN})}$ = 1.22V, I_{Q} = 20 μA , I_{SD} < 1 μA , ThinSOT Package
LT3009	20V, 20mA (I _{OUT}) Low Noise LDO	V_{IN} : 1.6V to 20V, $V_{OUT(MIN)}$ = 0.6V, I_Q = 3 μ A, I_{SD} < 1 μ A, SC70, 2mm \times 2mm DFN-6 Packages